#### Practice #5 - October 15, 2021

# **Ultra Wide Band Radio Fundamentals**

# **Spectral Characteristics of UWB Signals**

**DIET Department** 



## Outline

#### PPM analog modulator Deterministic modulation Random modulation

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**PPM-TH-UWB** 

## Outline

#### PPM analog modulator

Deterministic modulation Random modulation

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**PPM-TH-UWB** 

## PPM-TH similarity with analog theory

**1. PPM-TH-UWB**: if  $\theta_k$  is the dither process:

$$x(t) = \sum_{k \in \mathbb{Z}} p(t - kT_s - \theta_k)$$

**2. PPM-analog modulator**: modulation with m(t)

$$x(t) = \sum_{k \in \mathbb{Z}} p(t - kT_s - m(kT_s))$$

If max  $|m(t)| \le T_s/2$ , then the pulses are strictly non-overlapping.

- 1. Deterministic modulation.
- 2. Random modulation.

PPM analog modulator

L Deterministic modulation

### Deterministic modulation CHECKPOINT 3–1

**PPM-analog modulator**:

$$x(t) = \sum_{k \in \mathbb{Z}} p(t - kT_s - m(kT_s))$$

with modulating signal:

$$m(t) = A \cos(2\pi f_c t).$$

For the sake of simplicity, consider rect-pulse:

$$p(t) = \operatorname{rect}_{Tr}(t - T_r/2).$$

PPM analog modulator

Deterministic modulation

We shall generate the previous signal and evaluate its PSD.



The function declaration could be:

```
[signal,smp_time] = PPM_det(Ts,np,A,f0,Tr,pow,fs)
```

PPM analog modulator

Deterministic modulation

#### **Parameters:**

- > T<sub>s</sub> (mean pulse rep time): 2 [ns],
- > **np** (number of pulses): 10<sup>4</sup>,
- > Tr (rect pulse-width): 0.5 [ns],
- fs (sampling freq): 100 [GHz],
- > pow (avg signal power): -30 [dBm].

#### Two cases:

- 1. PPM-shift absence:
  - A (max PPM-shift): 0 [s],
  - f<sub>0</sub> (carrier freq): 0 [Hz].
- 2. PPM-shift presence:
  - A (max PPM-shift): 1 [ns],
  - f<sub>0</sub> (carrier freq): 50 [MHz].

L Deterministic modulation

## What to expect?

**PSD** explicit expression:

$$S_{XX}(f) = \frac{1}{T_s^2} \sum_{m \in \mathbb{Z}} \sum_{n \in \mathbb{Z}} \left| J_n(2\pi A f_{m,n}) \right|^2 \left| P(f_{m,n}) \right|^2 \delta(f - f_{m,n})$$

with  $f_{m,n} = m/Ts + nf_0$ .

Note that:  $J_n(x) \simeq 0$  if |n| > |x|, so  $|n| > 2\pi A \frac{m}{T_s}$  (so they are negligible in the summation)

Deterministic modulation

#### 1. PPM-shift absence:

-spectral-peaks: lines due to the absence of modulation,

- PSD-envelope: "sinc-like" (due to the pulse shape).



PPM analog modulator

Deterministic modulation

#### 2. PPM-shift presence:

- **spectral-peaks**: more lines, located @  $f = f_{m,n}$  and organized in **clusters** centered @  $m/T_s$ ,
- PSD-envelope: distortion of the "sinc-like" shape.



Deterministic modulation

What is the effect of the Bessel functions of first kind?

#### From **PSD** expression:

$$P_{x_{PPM}}(f) = \frac{1}{T_s^2} \sum_{m=-\infty}^{+\infty} \left| \sum_{n=-\infty}^{+\infty} \left| J_n \left( 2\pi A \left( \frac{m}{T_s} + nf_0 \right) \right) \right|^2 \left| P \left( \frac{m}{T_s} + nf_0 \right) \right|^2 \delta \left( f - \left( \frac{m}{T_s} + nf_0 \right) \right) \right|^2$$

Consider:

$$A = T_{s}/2, f_{0} = 1/(10T_{s})$$
$$J_{n}\left(2\pi \frac{T_{s}}{2}\left(\frac{m}{T_{s}} + \frac{n}{10T_{s}}\right)\right) = J_{n}\left(m\pi + \frac{n\pi}{10}\right)$$
$$A_{0}(n) = \left|J_{n}\left(\frac{n\pi}{10}\right)\right|^{2} \qquad A_{+1}(n) = \left|J_{n}\left(\pi + \frac{n\pi}{10}\right)\right|^{2} \qquad A_{-1}(n) = \left|J_{n}\left(-\pi + \frac{n\pi}{10}\right)\right|^{2}$$

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PPM analog modulator

Deterministic modulation

## Bessel order-zero cluster

 $(J_n \text{ contribution for } m = 0)$ 



PPM analog modulator

Deterministic modulation

#### Bessel first-order cluster

 $(J_n \text{ contribution for } m = \pm 1)$ 



PPM analog modulator

Deterministic modulation

# Bessel **second**-order cluster $(J_n \text{ contribution for } m = \pm 2)$



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PPM analog modulator

Random modulation

## Random modulation CHECKPOINT 3-3

**PPM-analog modulator:** 

$$x(t) = \sum_{k \in \mathbb{Z}} p(t - kT_s - m(kT_s))$$

where  $m[k] := m(kT_s)$  are IID samples of the SSS continuous stochastic process m(t).

We will choose  $m[k] \sim N(0, \sigma^2)$  and  $p(t) = rect_{Tr}(t - T_r/2)$ .

Random modulation

#### We shall generate the previous signal and evaluate its PSD.



The function declaration could be:

[signal,smp\_time] = PPM\_rndn(Ts,np,sigma,Tr,pow,fs).

PPM analog modulator

Random modulation

#### Parameters:

- T<sub>s</sub> (mean pulse rep time): 2 [ns],
- **np** (number of pulses): 10<sup>4</sup>,
- T<sub>r</sub> (rect pulse-width): 0.5 [ns],
- f<sub>s</sub> (sampling freq): 100 [GHz],
- **pow** (avg signal power): -30 [dBm].

We will choose: sigma (std dev of the gaussian) = 0.1 [ns].

Random modulation

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## What to expect?

**PSD** explicit expression:

$$S_{XX}(f) = \frac{|P(f)|^2}{T_s} \left[ 1 - |W(f)|^2 + \frac{|W(f)|^2}{T_s} \operatorname{comb}_{\frac{1}{T_s}}(f) \right]$$
  
ith  $\operatorname{comb}_T(t) := \sum_{n \in \mathbb{Z}} \delta(t - nT).$ 

Note the presence of both a continuous and a discrete component.

PPM analog modulator

Random modulation

**spectral-peaks: discrete term** predominant, due to a lack of asynchrony ( $\sigma$  too small with respect to pulse rep time),

**PSD-envelope:** "gaussian-like" (no longer "sinc-like"), due to  $|W(f)|^2$  (it determines the BW jointly with P(f)).



## Outline

#### PPM analog modulator Deterministic modulation Random modulation

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#### **PPM-TH-UWB**

PPM-TH-UWB CHECKPOINT 3-4 PPM-TH-UWB

$$x(t) = \sum_{k \in \mathbb{Z}} p(t - kT_s - \theta_k)$$

with dither process:

$$\theta_k = c_k T_c + a_k \varepsilon.$$

Depending on the values of  $\varepsilon$  and  $N_p$  (= code length), we can have three interesting cases:

- 1. no PPM-shift, no TH-code,
- 2. no PPM-shift, TH-code,
- 3. PPM, TH-code.

#### **Parameters:**

- nbits = 10; Ns = 5; Np = 5; Nh = 5; Tc = 1e-9; Tf = 5e-9; dPPM = 0.25e-9;Tm = 0.9e-9;powdBm = -30; fs = 30e9;
- % bitstream length
- % channel coder repetition factor
- % Time Hopping code length
- % code max value
- % Chip time
- % frame time
- % PPM shift
- % Pulse Duration [s]
- % avg Tx Power [dBm]
- % sampling frequency [GHz]

- 1. no PPM-shift ( $\varepsilon = 0$ ), no TH-code ( $N_p = 1$ ): periodic signal, so we expect spectral peaks @  $m/T_s$ .
  - spectral-peaks: lines located  $@ f = m/T_s$ ,
  - ▶ **PSD-envelope**: FT of 2<sup>nd</sup> derivative of a gaussian.



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- 2. no PPM-shift ( $\varepsilon = 0$ ), TH-code ( $N_p = N_s = 5$ ): spectral peaks @  $m/(N_s T_s) = m/T_b$  due to the  $(N_p = N_s)$ -induced periodicity, thus much closer spectral lines (*whitening* effect of the code)
  - spectral-peaks: lines located @  $f = m/T_b$ ,
  - PSD-envelope: FT of 2<sup>nd</sup> derivative of a gaussian.



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Even if the TH code reduces the number of peaks with high power contribution, it turns out that **it is not possible to remove all the peaks by only increasing the periodicity of the TH code.** Check the case : Ns = 5, Np = 5e3.



However, we can remove the strong peaks in the PSD by increasing  $N_h$  (max TH code value).



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#### 3. dPPM=0.25e-9 s, Np=Nh=5



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