

# Ultra Wide Band Radio Fundamentals

## Link Budget

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# Outline

Link budget in short

PSD and BW of interest

SNR and modulation

Simulation time!

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## Link budget in short

1. **What: Determine the maximum distance** of propagation at a given bit rate under a maximum BER constraint for the UWB point-to-point link.
  
2. **Why: Coexistence** with other radio signals requires compliance with emission masks regulated by the FCC.

## Link Budget Formula (1/2)

The BER constraint can be seen as an SNR constraint, thus:

$$\text{SNR}_0 \leq \text{SNR} = \frac{\mathcal{E}_r}{\mathcal{E}_N} = \frac{\mathcal{P}_r T_b}{\frac{1}{2} K T_0 F}$$

System margin

$$\text{SNR} / \text{SNR}_0 =: M \geq 1$$

$\mathcal{P}_r$  is the available received power

PSD of Tx signal

$$\mathcal{P}_r = \int_{f_L}^{f_H} \frac{2 \mathcal{S}_{pp}(f)}{A_{fs}(f)} df$$

$$A_{fs}(f) = \frac{4\pi D^2}{G_T G_R} \frac{4\pi}{\lambda^2}, \quad \lambda = c/f.$$

# Remind from lecture:

## 1. Required energy at the receiver

$$E_r = M \cdot SNR_{spec} \cdot E_{noise}$$

$$E_r = M \cdot SNR_{spec} \cdot \frac{1}{2} k (Temp_A + (F(f) - 1) Temp_0)$$

*receiving antenna temperature*  
*standard temperature (290K)*

$$P_r T_b = M \cdot SNR_{spec} \cdot \frac{1}{2} k (Temp_A + (F(f) - 1) Temp_0)$$

$$\rightarrow P_r = \frac{E_r}{T_b} = \frac{M SNR_0 E_N}{T_b}$$

## 2. Received power

$$P_r = 2 \int_{f_L}^{f_H} \frac{P_S(f)}{A_{FS}(f)} df = 2 \int_{f_L}^{f_H} \frac{P_S(f)}{(4\pi)^2 D^2 f^2} \frac{1}{G_T G_R c^2} df$$

$P_S(f)$  double-sided transmitted power spectral density

$A_{FS}(f)$  Free-space attenuation

## Link Budget Formula (2/2)

Solving in  $D$ :

$$D = \left[ \frac{G_T G_R \left(\frac{c}{4\pi}\right)^2 T_b}{M \text{SNR}_0 \mathcal{E}_N} \int_{f_L}^{f_H} \frac{2\mathcal{S}_{pp}(f)}{f^2} df \right]^{1/2}$$

Thus, the maximum link distance depends upon the:

- 1. bit rate:**  $T_b$  with  $D \propto 1/\sqrt{R_b}$
- 2. PSD of Tx signal and BW of interest:**  $\mathcal{S}_{pp}(f)$  and  $[f_L, f_H]$
- 3. min SNR to achieve the fixed BER:**  $\text{SNR}_0$

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## Choose the signal to transmit (1/2)

**CASE 1.** We transmit the 5<sup>th</sup> derivative of a gaussian pulse. You can accomplish this task in many ways:

1. writing the PSD directly\*, with  $N = 5$ :

$$\mathcal{S}_{pp}(f) = A_{max} \frac{(2\pi f \sigma)^{2N}}{N! e^{-N}} e^{-(2\pi f \sigma)^2}$$

where:

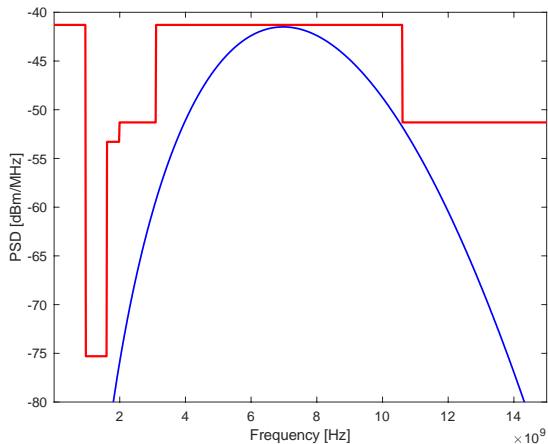
- $A_{max} = 10^{-13.15}$  W/Hz
- $\sigma = 51$  ps
- $f_{Max} = 30$  GHz

2. using the function `gaussian_wf` then computing the PSD.

\*[According to (Sheng et al., 2003)].

## Choose the signal to transmit (1/2)

Why the 5<sup>th</sup> derivative?



## Choose the signal to transmit (2/2)

**CASE 2.** We transmit an ideal signal that fully exploit the 3.1–10.6 GHz bandwidth of the FCC mask.

The PSD of the signal is the mask itself, in that range.

**NB:** verify that the total power of the signal is 0.55 mW.

## Choose the signal to transmit (2/2)

**CASE 2.** We transmit an ideal signal that takes full advantage of the 3.1–10.6 GHz bandwidth of the FCC mask.

Frequency in MHz	$EIRP_{mb}$ in dBm
0–960	-41.3
960–1610	-75.3
1610–1990	-53.3
1990–3100	-51.3
3100–10600	-41.3
Above 10600	-51.3

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# M-PAM and M-PPM probability of error in AWGN channel

## M-PAM

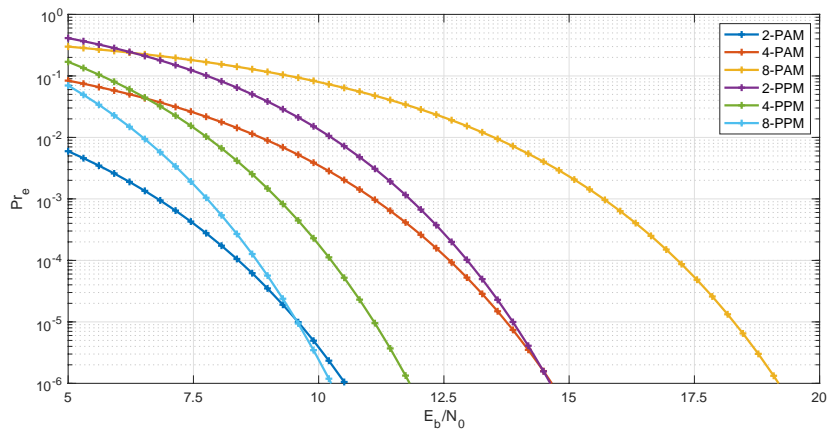
$$\Pr_e^{PAM} = \left(1 - \frac{1}{M}\right) \operatorname{erfc} y \quad \text{with} \quad y^2 = \frac{3 \log_2 M}{M^2 - 1} \frac{\mathcal{E}_b}{N_0}$$

## M-PPM

$$\Pr_e^{PPM} = \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{\infty} \left\{ 1 - \left[ 1 - \frac{1}{2} \operatorname{erfc} \left( \frac{\eta + m}{\sqrt{2}} \right) \right]^{M-1} \right\} e^{-\eta^2/2} d\eta$$

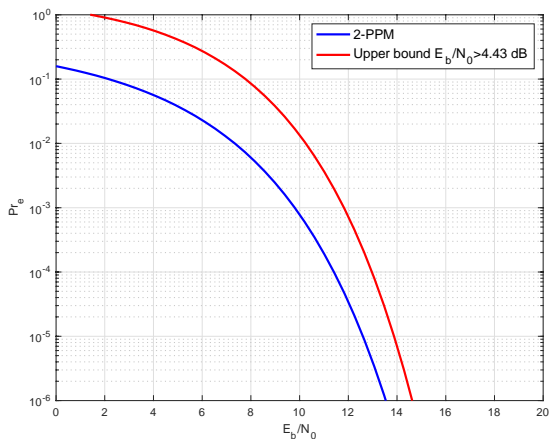
with  $m = \sqrt{2 \log_2 M} \frac{\mathcal{E}_b}{N_0}$ .

# M-PAM and M-PPM



# M-PPM approximation (tight for $E_b/N_0 > 4.43$ dB)

$$\Pr_e < e^{-\log_2 M (E_b/N_0 - 2 \log_e 2) / 2}$$





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## Simulation time! (1/3)

We shall write a script that, given a specified symbol  $Pr_e$ , returns the minimum required  $E_b/N_0$ .

The function declaration could be:

$$EbN0 = \text{getEbN0}(\text{MOD}, M, \text{Pre})$$

with:

- MOD (modulation type): 1 (PAM), 2 (PPM)
- M (bit per symbol)
- Pre (symbol probability of error):  $1e-3$
- Range of  $E_b/N_0$ : 0-20dB

## Simulation time! (2/3)

The core link-budget routine, that computes the maximum link distance, is the follow.

The function declaration could be:

```
D=distcmpt(EbN0min, Gt, Gr, MdB, FdB, Rb, PSDss,BOI)
```

with:

- EbN0min: required Eb/N0
- Gt: tx antenna gain
- Gr: rx antenna gain
- MdB: system margin in dB
- FdB: rx noise figure
- Rb: bit rate
- PSDss: single sided PSD
- BOI: bandwidth of Interest

## Simulation time! (3/3)

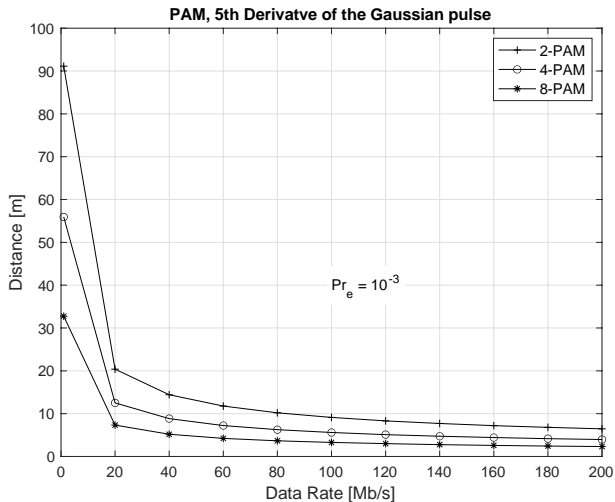
Parameters for function `distcmpt`:

$E_b N_0$ min: just computed	MdB = 5 dB
$G_t = 1$	$R_b = [1, 20:20:200]$ Mb/s
$G_r = 1$	$f_{MAX} = 30$ GHz
$F_{dB} = 7$ dB	

We will consider two cases for the bandwidth of interest and PSD :

- Case 1:
  - $PSD_{SS}$ : SS-PSD of 5<sup>th</sup> deriv. of Gaussian Pulse
  - BOI obtained as the -3 dB bandwidth of the pulse
- Case 2:
  - $PSD_{SS}$ : FCC mask
  - BOI = 3.1 – 10.6 GHz

# Results: case 1 - gaussian pulse, Mod = 1 (PAM)



# Results: case 2 - ideal signal, Mod = 1 (PAM)

