

Ultra Wide Band Radio Fundamentals

MUI models for IR-UWB

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Outline

SGA

- Sigma effect
- BER
- BER floor

PC

Theory vs. simulation

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Theory vs. simulation

Remind:

decision variable

$$Z = \int_{\hat{o}}^{N_s T_s + \hat{o}} r(t) m(t - \hat{o}) dt$$

$$Z = Z_u + \underbrace{Z_{MUI}}_{\text{the MUI term}} + Z_n$$

the MUI term

$$Z_{mui} = \sum_{i=1}^{N_i} \sqrt{E^{(i)}} \sum_{j=0}^{N_s} \int_{-\infty}^{\infty} p_0(t - \theta_j) (p_0(t) - p_0(t - \varepsilon)) dt$$

Z_{mui} is a random variable with mean zero and variance:

$$\sigma_{mui}^2 = \sum_{i=1}^{N_i} E^{(i)} N_s \frac{1}{T_s} \int_{-\infty}^{\infty} \left(\int_{-\infty}^{\infty} p_0(t - \tau) (p_0(t) - p_0(t - \varepsilon)) dt \right)^2 d\tau$$

σ_M^2 constant term which depends on both pulse shape and PPM shift

$$\sigma_{mui}^2 = \sum_{i=1}^{N_i} E^{(i)} N_s \frac{\sigma_M^2}{T_s} = \frac{N_s}{T_s} \sigma_M^2 \sum_{i=1}^{N_i} E^{(i)}$$

Sigma effect

The power of the MUI component of the decision variable is:

$$\sigma_{MUI}^2 = \frac{N_s}{T_s} \sigma_M^2 \sum_{i=1}^{N_i} E^{(i)}$$

If power control:

$$E^{(i)} = E_{RX} \forall i$$

where:

N_i – number of interferers

$E^{(i)}$ – energy of the i -th interferer

and:

$$\sigma_M^2 = \int_0^{T_s} \left\{ \int_0^{2T_M} p_0(t - \tau) [p_0(t) - p_0(t - \varepsilon)] dt \right\}^2 d\tau$$

If $\varepsilon > T_M$



$$\sigma_M^2 = \int_{-T_M}^{T_M} (R_0(\tau))^2 d\tau$$

Orthogonal pulses

Sigma effect: simulation

GOAL: show the effect of ε on σ_M^2 in PPM and PAM modulations

Write the function

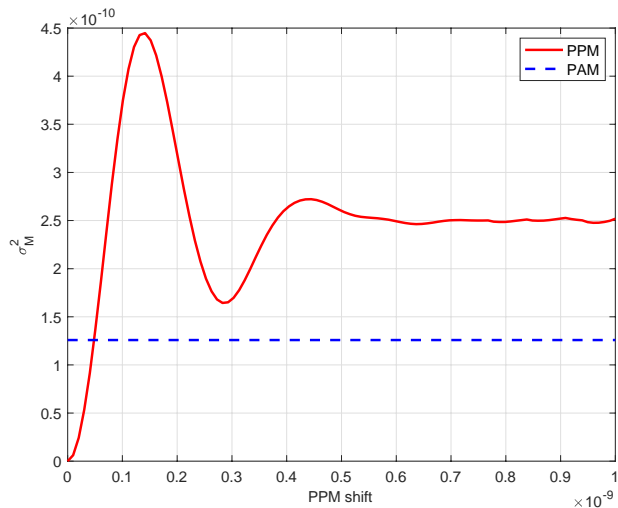
```
sm2 = sm2_mod(mod, pulse, dPPM, fc)
```

to highlight the *asymptotic* difference of σ_M^2 between PPM and PAM

Settings

- `mod`: 1 - PPM, 2 - PAM
- `pulse`: 2nd derivative of Gaussian pulse with $\tau=0.25$ ns and $T_M=1$ ns
- `dPPM`: PPM shift ε in $[0,1]$ ns
- `fc`: sampling frequency: 100 GHz

Sigma effect: result



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Remind: The Standard Gaussian Approximation

$$\Pr_b = \text{Prob}\left(N_S \sqrt{E_{RX}} (1 - R_0(\epsilon)) + Z_{mui} + Z_n < 0\right)$$

Gaussian random variable with mean zero and variance

$$\Pr_b = \frac{1}{2} \text{erfc}\left(\sqrt{\frac{1}{2} \text{SNR}_{spec}}\right)$$

$$\text{SNR}_{spec} = \frac{|Z_u|^2}{\sigma_{mui}^2 + \sigma_n^2} = \left(\left(\frac{|Z_u|^2}{\sigma_n^2} \right)^{-1} + \left(\frac{|Z_u|^2}{\sigma_{mui}^2} \right)^{-1} \right)^{-1} = \left((\text{SNR}_n)^{-1} + (\text{SNR}_{mui})^{-1} \right)^{-1}$$

$$\text{SNR}_n = \frac{N_S^2 E_{RX} (1 - R_0(\epsilon))^2}{N_S \mathcal{N}_0 (1 - R_0(\epsilon))} = N_S \frac{E_{RX}}{\mathcal{N}_0} (1 - R_0(\epsilon)) = \frac{E_b}{\mathcal{N}_0} (1 - R_0(\epsilon))$$

$$\text{SNR}_{mui} = \frac{N_S^2 E_{RX} (1 - R_0(\epsilon))^2}{\frac{N_S}{T_S} \sigma_M^2 \sum_{i=1}^{N_i} E^{(i)}} = \frac{T_S N_S (1 - R_0(\epsilon))^2}{\sigma_M^2 \sum_{i=1}^{N_i} \frac{E^{(i)}}{E_{RX}}} = \frac{(1 - R_0(\epsilon))^2}{R_b \sigma_M^2 \sum_{i=1}^{N_i} \frac{E^{(i)}}{E_{RX}}}$$

BER (SGA)

GOAL: evaluate the BER for PPM and PAM according to SGA for different numbers of interferers

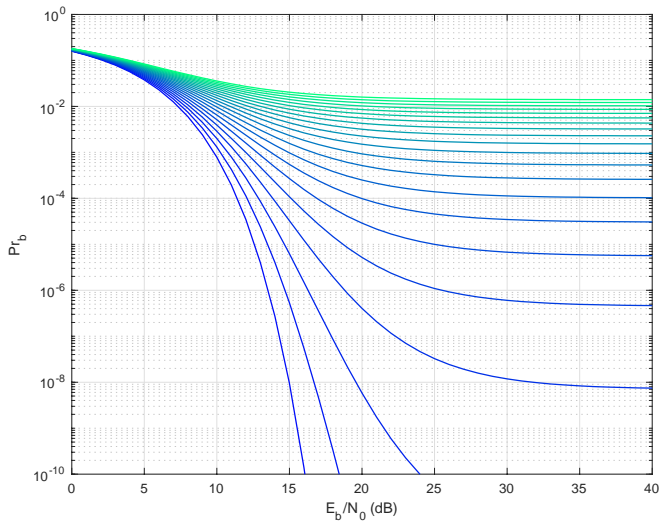
Write the function **MUIBER** defined as:

BER = **MUIBER**(mod, EbN0, erx0, erxMUI, pulse, Rb, dPPM, fc, gamma_r)

Settings

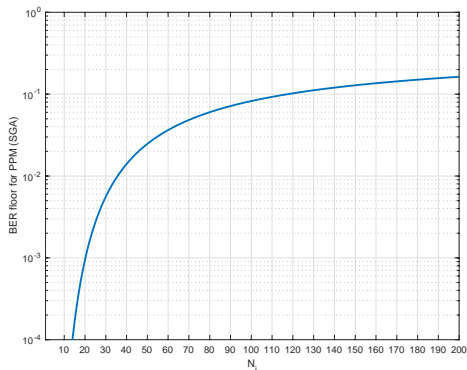
- mod: 1 - PPM, 2 - PAM
- EbN0: Eb/N0 in dB, [0, 40] dB
- erx0: normalized rx energy per pulse of the useful signal, 1 J
- erxMUI: vector of normalized rx energies per pulse of interferers, 1 J
- Rb: data rate, 20 Mb/s
- dPPM: 0.5 ns
- gamma_r: $\gamma_r = \frac{T_s}{T_b} N_s$, 1
- Ni (length of erxMUI): number of interferers, 1:2:39

Example: PPM



BER floor (SGA)

- Draw a plot showing the BER floor as a function of N_i for both modulation schemes, for $N_i=1:200$
- Example: PPM



- Compare the floor for PPM vs. PAM

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Reminder: Pulse Collision interference model

Assumptions:

- Asynchronous network \Rightarrow packet arrival is a Poisson process
- Pulse arrival is ALSO a Poisson process
- Collision probability is thus:

$$Pr_{PC} = 1 - e^{-2(N_u - 1) \frac{T_M}{T_s}}$$

- If pulse collision occurs then receiver decides randomly:

$$Pr_{PE} = 0.5 Pr_{PC}$$

- BER and PER (for a packet of length L, no error protection) are:

$$Pr_b = \sum_{i=\lceil \frac{N_s}{2} \rceil}^{N_s} \binom{N_s}{i} (Pr_{PE})^i (1 - Pr_{PE})^{N_s - i}$$

$$PER = 1 - (1 - Pr_b)^L$$

Reminder: Pulse Collision interference model

- Note that the PC model assumes that a collision occurs if two pulses overlap by even a small amount
- This may lead to an overestimation of the effect of interference
- This issue can be mitigated by introducing the concept of ***Effective Pulse Duration*** (EPD)
- The EPD is defined as the duration that contains a given percentage of the energy of the pulse
- EPD is then used in place of T_M in the estimation of the collision probability

Pulse Collision model: simulation time!

You will need to write two functions:

```
[epulse, EPD] = effpulse(pulse, fc, pE)
```

Returns the EPD that contains a percentage pE of the total energy of the pulse

```
BER_PC= prbcol(Nu, Ns, EPD, Ts)
```

Implements the BER formula recalled in the previous slide

Settings

Nu = 1:200

Tm = 1e-9

Ns = 5

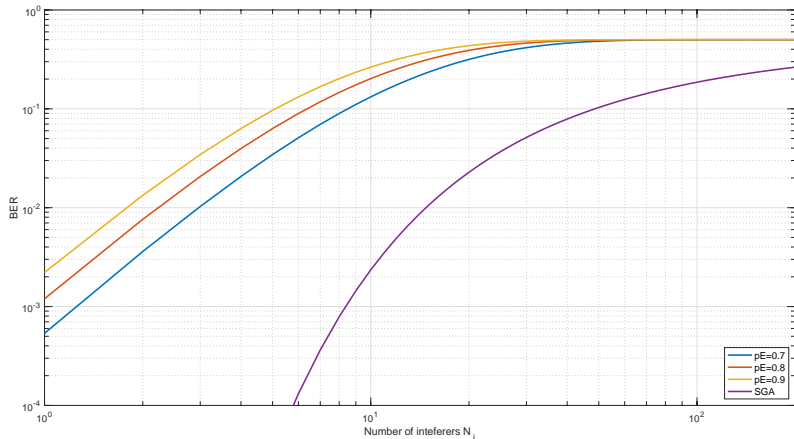
fc = 50e9

Ts = 3e-9

pE = 0.9, 0.8, 0.7

tau = 0.2e-9

Pulse Collision model: simulation time!



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BER: theory vs. simulation

GOAL: compare simulations with theory (SGA and PC models)

Procedure:

1. Generate the useful signal and N_i interfering signals with:

```
[bits, c, sTX, ref] = TX_BPPM_TH(nBits, Ns, fc, Tc, Ts, Np, Nh, dPPM, Tm, tau, powdBm);
```

2. Introduce asynchronism between users adding a random shift uniformly distributed in $[0, T_s]$ to each interfering user
3. Add noise to the signal obtained as the sum of useful and interfering signals using the `gnoise` function
4. Generate correlation mask and detect with the `corrmask` and `PPMreceiver` functions

Perform and average several runs in order to obtain reliable results

Settings

 $N_i = 5, 20, 60, 200$
 $\tau = 0.25e-9$
 $nBits = 1000$
 $powdBm = -30$
 $N_s = 3$
 $T_m = 0.5e-9$
 $dPPM = 0.5e-9$
 $N_h = 6$
 $T_s = 6e-9$
 $f_c = 50e9$
 $ebn0 = Inf$
 $N_p = 150$
 $T_c = 1e-9$
